## Super-resolution

## Optimization-Based Data Analysis

http://www.cims.nyu.edu/~cfgranda/pages/OBDA_spring16

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## Super-resolution

- Optics: Data-acquisition techniques to overcome the diffraction limit
- Image processing: Methods to upsample images onto a finer grid while preserving edges and hallucinating textures
- This lecture: Signal estimation from low-resolution data

Super-resolution of point sources
Spatial super-resolution
Spectral super-resolution
Deconvolution in reflection seismography
Conditioning of super-resolution
Linear methods
Periodogram
Local fitting
Parametric methods
Prony's method
Subspace methods
Matrix-pencil methods
Super-resolution via convex programming
Exact recovery
Super-resolution from noisy data

Super-resolution of point sources

## Spatial super-resolution

```
Spectral super-resolution
Deconvolution in reflection seismography
```

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## Limits of resolution in imaging

The resolving power of lenses, however perfect, is limited (Lord Rayleigh)

$\delta(t-\tau)$

optical system

$h(t-\tau)$

Diffraction imposes a fundamental limit on the resolution of optical systems

## Measurement model

Sensing mechanism acts as a low-pass filter

$$
\begin{aligned}
x_{\mathrm{LR}} & :=\phi * x \\
\widehat{x}_{\mathrm{LR}} & =\widehat{\phi} \widehat{x} \\
& =\widehat{\phi} \Pi_{\left[-f_{c}, f_{c}\right]} \widehat{x}
\end{aligned}
$$

High-frequency information is gone

We need prior assumptions to recover the signal

## Super-resolution of point sources



## Mathematical model

- Signal: Superposition of Dirac measures with support $T$

$$
x:=\sum_{j} c_{j} \delta_{t_{j}} \quad c_{j} \in \mathbb{C}, t_{j} \in T \subset[0,1]
$$

- Data: Convolution of signal and point-spread function

$$
\begin{aligned}
x_{\mathrm{LR}}(t) & :=\phi * x(t) \\
& =\sum_{t_{j} \in T} c_{j} \phi\left(t-t_{j}\right),
\end{aligned}
$$

Equivalently, low-pass Fourier coeffs with cut-off frequency $f_{c}$

$$
\begin{aligned}
y & =\mathcal{F}_{c} x \\
y_{k} & =\int_{0}^{1} e^{-i 2 \pi k t} x(\mathrm{~d} t)=\sum_{j} c_{j} e^{-i 2 \pi k t_{j}}, \quad k \in \mathbb{Z},|k| \leq f_{c}
\end{aligned}
$$

## Spatial Super-resolution

Spectrum

Signal


Data


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## Mathematical model

- Signal: Multisinusoidal signal

$$
\begin{gathered}
g(t):=\sum_{f_{j} \in T} c_{j} e^{-i 2 \pi f_{j} t} \\
\widehat{g}=\sum_{f_{j} \in T} c_{j} \delta_{f_{j}}
\end{gathered}
$$

- Data: $n$ samples measured at Nyquist rate

$$
g(k):=\sum_{f_{j} \in T} c_{j} e^{-i 2 \pi k f_{j}}, \quad 1 \leq k \leq n
$$

## Spectral Super-resolution

Spectrum

Signal


Data


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## Seismology



## Reflection seismology

Geological section
Acoustic impedance
Reflection coefficients


## Reflection seismology



Data $\approx$ convolution of pulse and reflection coefficients

## Sensing model for reflection seismology



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## Compressed sensing vs super-resolution

Estimation of sparse signals from undersampled measurements suggests connections to compressed sensing

Compressed sensing
Super-resolution

spectrum interpolation

spectrum extrapolation

## Compressed sensing




Spectrum of $x$

## Compressed sensing



Measurement operator $=$ random frequency samples

## Compressed sensing



## Compressed sensing



Aim: Study effect of measurement operator on sparse vectors

## Compressed sensing



Operator is well conditioned when acting upon any sparse signal (restricted isometry property)

## Compressed sensing



Operator is well conditioned when acting upon any sparse signal (restricted isometry property)

## Super-resolution



No discretization

## Super-resolution



Data: Low-pass Fourier coefficients

## Super-resolution



Data: Low-pass Fourier coefficients

## Super-resolution



Problem: If the support is clustered, the problem may be ill posed In super-resolution sparsity is not enough!

## Super-resolution



If the support is spread out, there is still hope We need conditions beyond sparsity

## Minimum separation

The minimum separation $\Delta$ of a discrete set $T$ is

$$
\Delta=\inf _{\left(t, t^{\prime}\right) \in T: t \neq t^{\prime}}\left|t-t^{\prime}\right|
$$



Example: 25 spikes, $f_{c}=10^{3}, \Delta=0.8 / f_{c}$

Signals


Data (in signal space)


Example: 25 spikes, $f_{c}=10^{3}, \Delta=0.8 / f_{c}$


Signals
Data (in signal space)

Example: 25 spikes, $f_{c}=10^{3}, \Delta=0.8 / f_{c}$

The difference is almost in the null space of the measurement operator


Difference


Spectrum

## Lower bound on $\Delta$

- Above what minimum distance $\Delta$ is the problem well posed?
- Numerical lower bound on $\Delta$ :

1. Compute singular values of restricted operator for different values of $\Delta_{\text {diff }}$
2. Find $\Delta_{\text {diff }}$ under which the restricted operator is ill conditioned
3. Then $\Delta \geq 2 \Delta_{\text {diff }}$


## Singular values of the restricted operator

Number of spikes $=s, f_{c}=10^{3}$



Phase transition at $\Delta_{\text {diff }}=1 / 2 f_{c} \rightarrow \Delta=1 / f_{c}$
Characterized asymptotically by Slepian's prolate spheroidal sequences

## Singular values of the restricted operator

Number of spikes $=s, f_{c}=10^{3}$



Phase transition at $\Delta_{\text {diff }}=1 / 2 f_{c} \rightarrow \Delta=1 / f_{c}$
Characterized asymptotically by Slepian's prolate spheroidal sequences

## Interpretation of $\lambda_{c}:=1 / f_{c}$

## Diameter of point-spread function


$\lambda_{c} / 2$ is the Rayleigh resolution distance

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## Periodogram

Spectrum of truncated data in spectral super-resolution

$$
\begin{aligned}
P(t) & =\mathcal{F}_{n}^{*} y \\
& =\sum_{t_{j} \in T} c_{j} D_{f_{c}}\left(t-t_{j}\right),
\end{aligned}
$$

$D_{f_{c}}$ is the periodized sinc or Dirichlet kernel

$$
D_{f_{c}}(t):=\sum_{k=-f_{c}}^{f_{c}} e^{i 2 \pi k t}= \begin{cases}1 & \text { if } t=0 \\ \frac{\sin \left(\left(2 f_{c}+1\right) \pi t\right)}{\left(2 f_{c}+1\right) \sin (\pi t)} & \text { otherwise }\end{cases}
$$

## Periodogram



## Windowing

Window function $\widehat{w} \in \mathbb{C}^{n}$

$$
\begin{aligned}
y_{\widehat{w}} & =y \cdot \widehat{w} \\
P_{\widehat{w}}(f) & =\mathcal{F}_{n}^{*} y_{\widehat{w}} \\
& =\sum_{t_{j} \in T} c_{j} w\left(t-t_{j}\right),
\end{aligned}
$$

## Windowing


-Window function
——Windowed data


## Windowing



Minimum separation: Periodogram

$$
\Delta=\frac{0.6}{f_{c}}
$$

$$
\Delta=\frac{1.2}{f_{c}}
$$



Minimum separation: Gaussian periodogram

$$
\Delta=\frac{0.6}{f_{c}}
$$

$$
\Delta=\frac{1.2}{f_{c}}
$$



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## Local fitting

Assume only one source

$$
x_{\mathrm{LR}}(t):=c_{1} \phi\left(t-t_{1}\right) .
$$

Estimation via best $\ell_{2}$-norm fit

$$
\begin{aligned}
t_{\text {est }} & =\arg \min _{\tilde{t}} \min _{\alpha \in \mathbb{C}} \mid\left\|x_{\mathrm{LR}}-\alpha \phi_{\tilde{t}}\right\|_{2} \\
& =\arg \max _{\tilde{t}}\left|\left\langle x_{\mathrm{LR}}, \phi_{\tilde{t}}\right\rangle\right|
\end{aligned}
$$

If sources are far we can compute local fits

Equivalent to matching pursuit

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## Prony polynomial

Signal

$$
x:=\sum_{j} c_{j} \delta_{t_{j}} \quad c_{j} \in \mathbb{C}, t_{j} \in T \subset[0,1]
$$

Data

$$
y_{k}:=\widehat{x}(k), \quad 0 \leq k \leq n-1 .
$$

Prony polynomial

$$
\begin{aligned}
P_{\text {prony }}(t) & :=\prod_{j=1}^{s}\left(1-e^{i 2 \pi\left(t-t_{j}\right)}\right) \\
& =1+\sum_{l=1}^{s} v_{l} e^{i 2 \pi / t}, \quad v_{0}:=1
\end{aligned}
$$

## Prony polynomial



## Computing the Prony polynomial

By construction

$$
\left\langle P_{\text {prony }}, x\right\rangle=0
$$

By Parseval's Theorem

$$
\begin{aligned}
\left\langle P_{\text {prony }}, x\right\rangle & =\langle v, \widehat{x}\rangle \\
& =\sum_{k=0}^{s} v_{k} \overline{y_{k}} \quad \text { if } s+1 \leq n
\end{aligned}
$$

## Computing the Prony polynomial

By construction

$$
\left\langle P_{\text {prony }}, e^{2 \pi k^{\prime} t} x\right\rangle=0
$$

By Parseval's Theorem

$$
\begin{aligned}
\left\langle P_{\text {prony }}, e^{2 \pi k^{\prime} t} x\right\rangle & =\left\langle v, \widehat{x}_{k^{\prime}}\right\rangle \\
& =\sum_{k=0}^{s} v_{k} \overline{y_{k+k^{\prime}}} \quad \text { if } s+k^{\prime} \leq n-1
\end{aligned}
$$

## Prony's method

1. Form the system of equations

$$
\left[\begin{array}{cccc}
y_{1} & y_{2} & \ldots & y_{s} \\
y_{2} & y_{3} & \ldots & y_{s+1} \\
\ldots & \ldots & \ldots & \ldots \\
y_{s} & y_{s+1} & \ldots & y_{n-1}
\end{array}\right]\left[\begin{array}{c}
\tilde{v}_{1} \\
\tilde{v}_{2} \\
\cdots \\
\tilde{v}_{s}
\end{array}\right]=-\left[\begin{array}{c}
y_{0} \\
y_{1} \\
\ldots \\
y_{s-1}
\end{array}\right]
$$

2. Solve the system and set $\tilde{v}_{0}=1$
3. Roots of polynomial with coeffs $\tilde{v}_{0}, \ldots, \tilde{v}_{s}: z_{1}, \ldots, z_{s}$
4. For $z_{j}=e^{i 2 \pi \tau}$ include $\tau$ in estimated support

## Prony's method

$$
\begin{aligned}
{\left[\begin{array}{cccc}
y_{1} & y_{2} & \ldots & y_{s} \\
y_{2} & y_{3} & \ldots & y_{s+1} \\
\ldots & \ldots & \ldots & \ldots \\
y_{s} & y_{s+1} & \ldots & y_{n-1}
\end{array}\right]=} & {\left[\begin{array}{cccc}
e^{-i 2 \pi t_{1}} & e^{-i 2 \pi t_{2}} & \ldots & e^{-i 2 \pi t_{s}} \\
e^{-i 2 \pi 2 t_{1}} & e^{-i 2 \pi 2 t_{2}} & \ldots & e^{-i 2 \pi 2 t_{s}} \\
\ldots & \ldots & \ldots & \ldots \\
e^{-i 2 \pi s t_{1}} & e^{-i 2 \pi s t_{2}} & \ldots & e^{-i 2 \pi s t_{s}}
\end{array}\right] } \\
& {\left[\begin{array}{cccc}
c_{1} & 0 & \ldots & 0 \\
0 & c_{2} & \ldots & 0 \\
\ldots & \ldots & \ldots & \ldots \\
0 & 0 & \ldots & c_{s}
\end{array}\right] } \\
& {\left[\begin{array}{cccc}
1 & e^{-i 2 \pi t_{1}} & \ldots & e^{-i 2 \pi(s-1) t_{1}} \\
1 & e^{-i 2 \pi t_{2}} & \ldots & e^{-i 2 \pi(s-1) t_{2}} \\
\ldots & \ldots & \ldots & \ldots \\
1 & e^{-i 2 \pi t_{s}} & \ldots & e^{-i 2 \pi(s-1) t_{s}}
\end{array}\right] }
\end{aligned}
$$

## Vandermonde matrix

For any distinct $s$ nonzero $z_{1}, z_{2}, \ldots, z_{s} \in \mathbb{C}$ and any $m_{1}, m_{2}, s$ such that $m_{2}-m_{1}+1 \geq s$

$$
\left[\begin{array}{cccc}
z_{1}^{m_{1}} & z_{2}^{m_{1}} & \cdots & z_{s}^{m_{1}} \\
z_{1}^{m_{1}+1} & z_{2}^{m_{1}+1} & \cdots & z_{s}^{m_{1}+1} \\
z_{1}^{m_{1}+2} & z_{2}^{m_{1}+2} & \cdots & z_{s}^{m_{1}+2} \\
& & \cdots & \\
z_{1}^{m_{2}} & z_{2}^{m_{2}} & \cdots & z_{s}^{m_{2}}
\end{array}\right]
$$

is full rank

## No noise


$\mathrm{SNR}=140 \mathrm{~dB}\left(\right.$ relative $\ell_{2}$ norm of noise $\left.=10^{-8}\right)$


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## Alternative interpretation of Prony's method

Prony's method finds nonzero vector in the null space of $Y(s+1)^{T}$

$$
Y(m):=\left[\begin{array}{cccc}
y_{0} & y_{1} & \cdots & y_{n-m} \\
y_{1} & y_{2} & \cdots & y_{n-m+1} \\
\cdots & \cdots & \cdots & \cdots \\
y_{m-1} & y_{m} & \cdots & y_{n-1}
\end{array}\right]
$$

The vector corresponds to the coefficients of the Prony polynomial

## Notation: Sinusoidal atoms

For $k>0$

$$
a_{0: k}(t):=\left[\begin{array}{c}
1 \\
e^{-i 2 \pi t} \\
e^{-i 2 \pi 2 t} \\
\cdots \\
e^{-i 2 \pi k t}
\end{array}\right]
$$

$$
A_{0: k}(T):=\left[\begin{array}{llll}
a_{0: k}\left(t_{1}\right) & a_{0: k}\left(t_{2}\right) & \cdots & a_{0: k}\left(t_{s}\right)
\end{array}\right]
$$

## Decomposition

$$
\begin{aligned}
Y(m) & =\left[\begin{array}{llll}
a_{0: m-1}\left(t_{1}\right) & a_{0: m-1}\left(t_{2}\right) & \cdots & a_{0: m-1}\left(t_{s}\right)
\end{array}\right] \\
& {\left[\begin{array}{cccc}
c_{1} & 0 & \cdots & 0 \\
0 & c_{2} & \cdots & 0 \\
\cdots & \cdots & \cdots & \cdots \\
0 & 0 & \cdots & c_{s}
\end{array}\right]\left[\begin{array}{c}
a_{0: n-m}\left(t_{1}\right)^{T} \\
a_{0: n-m}\left(t_{2}\right)^{T} \\
\cdots \\
a_{0: n-m}\left(t_{s}\right)^{T}
\end{array}\right] } \\
& =A_{0: m-1}(T) \subset A_{0: m}(T)^{T}
\end{aligned}
$$

Idea: To estimate $T$ find $a_{0: m-1}(t)$ in the column space of $Y(m)$

## Pseudospectrum

To find atoms that are close to the column space of $Y(m)$

- Compute orthogonal complement $\mathcal{N}$ of column space of $Y(m)$
- Locate local maxima of pseudospectrum

$$
P_{\mathcal{N}}(t)=\log \frac{1}{\left|\mathcal{P}_{\mathcal{N}}\left(a_{0: m-1}(t)\right)\right|^{2}}
$$

## Empirical covariance matrix

$\mathcal{N}$ is the null space of the empirical covariance matrix

$$
\begin{aligned}
\Sigma(m) & =\frac{1}{n-m+1} Y Y^{*} \\
& =\frac{1}{n-m+1} \sum_{j=0}^{n-m}\left[\begin{array}{c}
y_{j} \\
y_{j+1} \\
\ldots \\
y_{j+m-1}
\end{array}\right]\left[\begin{array}{llll}
\overline{y_{j}} & \overline{y_{j+1}} & \ldots & \overline{y_{j+m-1}}
\end{array}\right]
\end{aligned}
$$

## Pseudospectrum

$$
Y(m)=A_{0: m-1}(T) C A_{0: m}(T)^{T}
$$

implies

$$
\begin{array}{ll}
P_{\mathcal{N}}\left(t_{j}\right)=\infty, & \text { for } t_{j} \in T \\
P_{\mathcal{N}}(t)<\infty, & \text { for } t \notin T
\end{array}
$$

## Pseudospectrum: No noise



## Pseudospectrum: SNR $=140 \mathrm{~dB}, n=2 \mathrm{~s}$



## Multiple-signal classification (MUSIC)

1. Build the empirical covariance matrix $\Sigma(m)$
2. Compute the eigendecomposition of $\Sigma(m)$
3. Select $U_{\mathcal{N}}$ corresponding to $m-s$ smallest eigenvalues
4. Estimate support by computing the pseudospectrum

## Pseudospectrum: $\mathrm{SNR}=40 \mathrm{~dB}, n=81, m=30$



## Pseudospectrum: SNR $=1 \mathrm{~dB}, n=81, m=30$



## Probabilistic model: Signal

$$
x=\sum_{t_{j} \in T} c_{j} \delta_{t_{j}}=\sum_{t_{j} \in T}\left|c_{j}\right| e^{i \phi_{j}} \delta_{t_{j}}
$$

The phases $\phi_{j}$ are independent and uniformly distributed in $[0,2 \pi]$

$$
\begin{aligned}
& \mathrm{E}(x)=0 \\
& \mathrm{E}\left[c c^{*}\right]=D_{c}:=\left[\begin{array}{cccc}
\left|c_{1}\right|^{2} & 0 & \ldots & 0 \\
0 & \left|c_{2}\right|^{2} & \ldots & 0 \\
\ldots & \ldots & \ldots & \ldots \\
0 & 0 & \ldots & \left|c_{s}\right|^{2}
\end{array}\right]
\end{aligned}
$$

## Probabilistic model: Noise

Noise $z$ is a zero-mean Gaussian vector with covariance $\sigma^{2} /$

$$
\begin{aligned}
\tilde{y}_{k} & :=\int_{0}^{1} e^{-i 2 \pi k t} x(\mathrm{~d} t)+z_{k} \\
& =\sum_{t_{j} \in T} c_{j} e^{-i 2 \pi k t_{j}}+z_{k} \\
\tilde{y} & =A_{0: m-1}(T) c+z
\end{aligned}
$$

Covariance matrix of the data

$$
\mathrm{E}\left[\tilde{y} \tilde{y}^{*}\right]=A_{1: m} D_{c} A_{1: m}^{*}+\sigma^{2} I
$$

## Eigendecomposition of covariance matrix

Eigendecomposition of $\mathrm{E}\left[\tilde{y}^{\tilde{y}} \tilde{}^{*}\right]$

$$
\mathrm{E}\left[\tilde{y} \tilde{y}^{*}\right]=\left[\begin{array}{ll}
U_{\mathcal{S}} & U_{\mathcal{N}}
\end{array}\right]\left[\begin{array}{cc}
\Lambda+\sigma^{2} I_{s} & 0 \\
0 & \sigma^{2} I_{n-s}
\end{array}\right]\left[\begin{array}{c}
U_{\mathcal{S}}^{*} \\
U_{\mathcal{N}}^{*}
\end{array}\right],
$$

- $U_{\mathcal{S}} \in \mathbb{C}^{m \times s}$ : unitary matrix that spans column space of $A_{1: m}$
- $U_{\mathcal{N}} \in \mathbb{C}^{m \times(m-s)}$ : unitary matrix spanning the orthogonal complement
- $\Lambda \in \mathbb{C}^{k \times k}$ is a diagonal matrix with positive entries
$\Delta=\frac{0.6}{f_{c}}, \mathrm{SNR}=20 \mathrm{~dB}, n=81, m=40$



## $\Delta=\frac{1.2}{f_{c}}, \mathrm{SNR}=20 \mathrm{~dB}, n=81, m=40$



## Different values of $m$

$\mathrm{SNR}=61 \mathrm{~dB}$


## Different values of $m$

$$
\mathrm{SNR}=21 \mathrm{~dB}
$$



## Different values of $m$

$\mathrm{SNR}=1 \mathrm{~dB}$


Eigenvalues


Eigenvalues


Eigenvalues


Wrong $s(s-1)$

SNR $=21 \mathrm{~dB}$


## Wrong $s(s+1)$

$\mathrm{SNR}=21 \mathrm{~dB}$


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## Low-rank model

$$
\begin{aligned}
Y_{0} & =\left[\begin{array}{cccc}
y_{0} & y_{1} & \cdots & y_{n-m} \\
y_{1} & y_{2} & \cdots & y_{n-m+1} \\
\cdots & \cdots & \cdots & \cdots \\
y_{m-1} & y_{m} & \cdots & y_{n-1}
\end{array}\right] \\
& =A_{0: m-1}(T) C A_{0: n-m}(T)^{T} \\
& =\sum_{t_{j} \in T} c_{j} a_{0: m-1}\left(t_{j}\right) a_{0: n-m}\left(t_{j}\right)^{T}
\end{aligned}
$$

## Matrix pencil

The matrix pencil of two matrices $M_{1}, M_{2}$ is

$$
L_{M_{1}, M_{2}}(\mu):=M_{2}-\mu M_{1}, \quad \mu \in \mathbb{C}
$$

The set of rank-reducing values $\mathcal{R}$ of a matrix pencil satisfy

$$
\operatorname{rank}\left(L_{M_{1}, M_{2}}(\mu)\right)=\operatorname{rank}\left(L_{M_{1}, M_{2}}\left(\mu_{j}\right)\right)+1
$$

for all $\mu_{j} \in \mathcal{R}$ and $\mu \notin \mathcal{R}$

## Matrix pencil

We consider the matrix-pencil of $Y_{0}$ and

$$
\begin{aligned}
Y_{1} & =\left[\begin{array}{cccc}
y_{1} & y_{2} & \cdots & y_{n-m+1} \\
y_{2} & y_{3} & \cdots & y_{n-m+2} \\
\cdots & \cdots & \cdots & \cdots \\
y_{m} & y_{m+1} & \cdots & y_{n}
\end{array}\right] \\
& =A_{1: m}(T) C A_{0: n-m}(T)^{T} \\
& =\sum_{t_{j} \in T} c_{j} a_{1: m}\left(t_{j}\right) a_{0: n-m}\left(t_{j}\right)^{T}
\end{aligned}
$$

$\exp (i 2 \pi \tau)$ is a rank-reducing value of $L_{Y_{0}, Y_{1}}$ if and only if $\tau \in T$

## Computing the rank-reducing values

Let $Y_{0}=U_{0} \Sigma_{0} V_{0}^{*}$ be the singular-value decomposition of $Y_{0}$

The $s$ eigenvalues of the matrix

$$
M=V_{0} \Sigma_{0}^{-1} U_{0}^{*} Y_{1}
$$

are equal to $\exp \left(i 2 \pi t_{j}\right)$ for $1 \leq j \leq s$

## Proof

$$
a_{1: m}(\tau)=\exp (i 2 \pi \tau) a_{0: m-1}(\tau)
$$

$$
A_{0: m-1}(T)=A_{0: m-1}(T) \Phi
$$

$$
\Phi:=\left[\begin{array}{cccc}
e^{i 2 \pi t_{1}} & 0 & \ldots & 0 \\
0 & e^{i 2 \pi t_{2}} & \ldots & 0 \\
\ldots & \ldots & \ldots & \ldots \\
0 & 0 & \ldots & e^{i 2 \pi t_{s}}
\end{array}\right]
$$

## Proof

$$
\begin{aligned}
& Y_{0}=A_{0: m-1}(T) C A_{0: n-m}(T)^{T} \\
& C A_{0: n-m}(T)^{T}=U \Sigma V^{*} \\
& C A_{0: n-m}(T)^{T} V \Sigma^{-1} U^{*}=1 \\
& Y_{1}=A_{1: m}(T) C A_{0: n-m}(T)^{T} \\
& \quad=A_{0: m-1}(T) \Phi C A_{0: n-m}(T)^{T} \\
& =A_{0: m-1}(T) C A_{0: n-m}(T)^{T} V \Sigma^{-1} U^{*} \Phi C A_{0: n-m}(T)^{T} \\
& =Y_{0} V \Sigma^{-1} U^{*} \Phi U \Sigma V^{*}
\end{aligned}
$$

## Proof

$$
\begin{aligned}
& V_{0} \Sigma_{0}^{-1} U_{0}^{*} Y_{0} V
\end{aligned}=V_{0} V_{0}^{*} V=V ~ \begin{aligned}
& V_{0} \Sigma_{0}^{-1} U_{0}^{*} Y_{1}=V_{0} \Sigma_{0}^{-1} U_{0}^{*} Y_{0} V \Sigma^{-1} U^{*} \Phi U \Sigma V^{*} \\
&=V \Sigma^{-1} U^{*} \Phi U \Sigma V^{*} \\
&=P^{-1}\left[\begin{array}{ll}
\Phi & 0 \\
0 & 0
\end{array}\right] P \\
& P:=\left[\begin{array}{ll}
U & 0 \\
0 & 1
\end{array}\right]\left[\begin{array}{ll}
\Sigma & 0 \\
0 & 1
\end{array}\right]\left[\begin{array}{l}
V^{*} \\
V_{\perp}^{*}
\end{array}\right]
\end{aligned}
$$

## Spectral super-resolution via matrix pencil

1. Build $Y_{0}=U_{0} \Sigma_{0} V_{0}^{*}$ and $Y_{1}$
2. Compute the $s$ largest eigenvalues $\lambda_{1}, \lambda_{2}, \ldots, \lambda_{s}$ of $V_{0} \Sigma_{0}^{-1} U_{0}^{*} Y_{1}$
3. Output the phase of $\lambda_{1}, \lambda_{2}, \ldots, \lambda_{s}$ divided by $2 \pi$

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## Sensing model for reflection seismology



## Suggestion of various geophysicists: Minimize $\ell_{1}$ norm

# Deconvolution with the $\boldsymbol{\ell}_{1}$ norm <br> Howard L. Taylor,* Stephen C. Banks, ${ }^{\ddagger}$ and John F. McCoy ${ }^{5}$ <br> LINEAR INVERSION OF BAND-LIMITED REFLECTION SEISMOGRAMS* 

FADIL SANTOSA $\dagger$ AND WILLIAM W. SYMES $\ddagger$

GEOPHYSICS, VOL. 44, NO. 1 (JANUARY 1979)

SIAM J. SCI. Stat. COMPUT.
Vol. 7, No. 4, October 1986

Reconstruction of a sparse spike train from a portion of its spectrum and application to high-resolution deconvolution

Shlomo Levy* and Peter K. Fullagar $\ddagger$

ROBUST MODELING WITH ERRATIC DATA $\dagger$

GEOPHYSICS, VOL. 46, NO. 9 (SEPTEMBER 1981)

## Minimum $\ell_{1}$-norm estimate

```
minimize
subject to estimate * pulse = data
|estimate||
```

Reflection coefficients


Estimate


## Total-variation norm

- Continuous counterpart of the $\ell_{1}$ norm
- If $x=\sum_{j} c_{j} \delta_{t_{j}}$ then $\|x\|_{T V}=\sum_{j}\left|c_{j}\right|$
- Not the total variation of a piecewise-constant function
- Formal definition: For a complex measure $\nu$

$$
\|\nu\|_{\mathrm{TV}}=\sup _{\|f\|_{\infty} \leq 1, f \in C(\mathbb{T})} \int_{\mathbb{T}} \overline{f(t)} x(\mathrm{~d} t)
$$

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## Super-resolution via convex programming

For data of the form $y=\mathcal{F}_{c} x$, we solve

$$
\min _{\tilde{x}}\|\tilde{x}\|_{T V} \quad \text { subject to } \quad \mathcal{F}_{c} \tilde{x}=y
$$

over all finite complex measures $\tilde{x}$ supported on $[0,1]$

Exact recovery is guaranteed if $\Delta \geq \frac{1.26}{f_{c}}$

## Dual certificate

The same as for the $\ell_{1}$ norm, but now $q$ is a function

$$
\begin{array}{ll}
q:=\mathcal{F}_{c}^{*} v & \\
q_{i}=\operatorname{sign}\left(x_{i}\right) & \text { if } x_{i} \neq 0 \\
\left|q_{i}\right|<1 & \text { if } x_{i}=0
\end{array}
$$

The rows of $\mathcal{F}_{c}$ are low pass sinusoids instead of random sinusoids

## Certificate for super-resolution



Aim: Interpolate sign pattern

## Certificate for super-resolution



1st idea: Interpolation with a low-frequency fast-decaying kernel $K$

$$
c(t)=\sum_{i: x_{i} \neq 0} \alpha_{i} K(t-i)
$$

## Certificate for super-resolution



1st idea: Interpolation with a low-frequency fast-decaying kernel $K$

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## Certificate for super-resolution



Problem: Magnitude of certificate locally exceeds 1

## Certificate for super-resolution



Problem: Magnitude of certificate locally exceeds 1
Solution: Add correction term and force the derivative of the certificate to equal zero on the support

$$
c(t)=\sum_{i: x_{i} \neq 0} \alpha_{i} K(t-i)+\beta_{i} K^{\prime}(t-i)
$$

## Certificate for super-resolution



Problem: Magnitude of certificate locally exceeds 1
Solution: Add correction term and force the derivative of the certificate to equal zero on the support

$$
c(t)=\sum_{i: x_{i} \neq 0} \alpha_{i} K(t-i)+\beta_{i} K^{\prime}(t-i)
$$

## Certificate for super-resolution



Similar construction works for bandpass point-spread functions relevant to reflection seismology

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## Super-resolution from noisy data

Additive-noise model

$$
y=\mathcal{F}_{n} x+z
$$

Relaxed optimization problem

$$
\min _{\tilde{x}}\|\tilde{x}\|_{\text {TV }} \quad \text { subject to } \quad\left\|\mathcal{F}_{n} \tilde{x}-y\right\|_{2}^{2} \leq \delta
$$

$\delta$ is an estimate of the noise level

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Relaxed optimization problem

$$
\min _{\tilde{x}}\|\tilde{x}\|_{\text {TV }} \quad \text { subject to } \quad\left\|\mathcal{F}_{n} \tilde{x}-y\right\|_{2}^{2} \leq \delta
$$

$\delta$ is an estimate of the noise level
$\Delta=\frac{0.6}{f_{c}}, \mathrm{SNR}=20 \mathrm{~dB}, f_{c}=40$


## $\Delta=\frac{1.2}{f_{c}}, \mathrm{SNR}=20 \mathrm{~dB}, f_{c}=40$



## Deconvolution with the $\ell_{1}$ norm (Taylor, Banks, McCoy '79)



